

ASYMMETRICAL LASER COMMUNICATION TRANSCEIVER CONFIGURATION

CROSS REFERENCE TO RELATED APPLICATIONS

The present application claims under 35 U.S.C. § 119(e) the benefits of U.S. Provisional Application Serial No. 60/264,079, filed January 24, 2001, entitled "On Axis Laser Receiver Wavelength Demultiplexer with Integral Immersion Lensed Detectors", which is incorporated herein by reference. Cross reference is made to U.S. Patent Applications Serial No. 09/339,316, filed June 23, 1999, to Smith entitled "RECEIVING MULTIPLE WAVELENGTHS AT HIGH TRANSMISSION RATES", which is incorporated herein by this reference.

FIELD OF THE INVENTION

The present invention relates to laser communication, and more specifically to an apparatus and method to transmit through free space and receive high data rate signals.

BACKGROUND INFORMATION

Various communication systems are known for transmitting through the atmosphere. Most commonly, microwave communication devices are used for this communication. Additionally, various optical techniques for communication are known.

Microwave technologies used for data links through the atmosphere generally suffer from low data rates. For example, an 11 to 14 or 20 to 30 gigahertz (GHz) carrier is sometimes used to transmit between a satellite and a ground station. However, modulating 1 gigabit/sec. data rates on the 18 GHz carrier is not currently practically accomplished. By

increasing the carrier frequency, the data rate may also be increased. For example, a 60 GHz carrier could support modulation data rates of 1 gigabit/sec. or higher. However, the atmosphere itself attenuates the 60 GHz carrier such that transmission through the atmosphere is not practical.

5 Terrestrial-based free space optical transmission systems are being considered as a replacement for free space microwave communication systems. Laser communication systems theoretically can transmit through the atmosphere at high data rates. Laser communication systems must be properly designed to minimize the risk of high bit error rates (BER) due to atmospheric disturbances from thermal air currents and inclement weather conditions.

10 The effects of atmospheric disturbances from thermal air currents along the optical path are typically greatest at the transmitter and receiver. For a terrestrial link, the transmitter and receiver are typically (compared to other parts of the optical path) in closest proximity to thermal sources, namely solid surfaces such as rooftops, which absorb solar flux and warm the air. The warm air rises through cooler air thereby generating turbulence and providing a non-uniform propagation medium for the optical signal. At the transmitter, such disturbances can cause different portions of the optical beam to pass through differently moving air currents, causing the propagating beam to break up, so leaving the receive aperture in an intensity minimum. Similar disturbances near the receive aperture can cause the focal spot to break up, leaving the detector in an intensity minimum. Each of these distinct processes can produce burst errors in the communicated data and must be addressed separately.

Space-based optical communication links between widely separated laser transceivers are generally designed to minimize the pointing criticality, aperture diameters for a given link range, laser power and data rate, by use of matched, common transmitter and receiver apertures. This symmetrical link geometry serves, in a homogeneous medium such as free space, to provide the maximum power transfer for the smallest terminal mass and volume. When the cost of dichroically separating the transmit and receive paths within the transceiver is prohibitive, a smaller transmit aperture is commonly located near a large receive aperture at each transceiver. Such an asymmetrical transceiver design is particularly common for atmospheric laser links to minimize the detrimental effect of atmospheric disturbances on BER. Even though a smaller transmit aperture can reduce interference, beams can still wander off or destructively interfere at the intended receiver. To address this problem, multiple (typically four) redundant transmitters are often used to reduce the probability of deep fades in the received power. This arrangement can still be inadequate, particularly in the afternoon when solar-induced thermal currents are most severe.

SUMMARY OF THE INVENTION

In accordance with the present invention, apparatuses and methods for high-bandwidth wireless data transmission, particularly through the atmosphere, are disclosed. Typically, the effects of atmospheric turbulence at the transmitter are maintained at or below predetermined or acceptable levels by using a (preferably collimated) beam having a smaller diameter than the inner scale as measured at the transmit aperture (*e.g.*, a correspondingly high divergence beam) As used herein, "inner scale" refers to the smallest distance across

which an air temperature difference can be sustained, "divergence" to the ratio of beam diameter to propagation distance. Such beams can substantially minimize power fades from atmospheric turbulence and therefore the occurrence of burst errors.

By keeping the critical first few meters of especially strong turbulence within the Rayleigh range (i.e., given by $\frac{(beam\ diameter)^2}{1.27\lambda}$), the initial beam diameter is preferably

maintained small enough as to see only minimal refractive index variation across its diameter, thereby avoiding the wavefront distortions which could have Fourier-transformed to an interference pattern in the far-field. The beam thus inhibits large area destructive interference near the receive aperture. The far-field divergence angle is large enough (typically from about 0.1 to about 2.0 milliradians and more typically from about 0.4 to about 1.0 milliradian) that it cannot be steered off the receiver by the turbulent air. The bigger far-field divergence angle reduces the interference of portions of the beam beyond its Rayleigh range, by virtue of their divergent Poynting vectors. In one configuration, a relatively large, low quality laser communication receiver aperture (i.e., photon bucket), typically on the order of at least about 0.2m and more typically about 0.4m depending on link distance, can then adjust for the big transmit divergence by subtending a large solid angle from the transmitter. The larger aperture captures enough of the radiation to compensate for the lower received radiation intensity from the purposely broadened transmit beam.

The transmitter configuration may be combined with an optical receiver having a primary optical lens that is a diffractive optical element, such as a transmissive or reflective

hologram, zone plate, *etc.*, that receives radiation modulated by high frequency data. As will be appreciated, a diffractive optical element spatially separates different wavelengths of radiation or wavelength bands, preferably focusing each wavelength or wavelength band of radiation to a different spatial location. A detector assembly, that is responsive to the
5 respective focused radiation at the detector, is preferably positioned at or near each focal area or point. Each detector assembly detects the data in the respective beam focused on the detector assembly. In one configuration, the detector assemblies are in a stacked or end-to-end configuration and are located along a common axis, which can be the optical axis of the primary optical lens.

10 This receiver configuration can realize high data transmission rates through free space without the drawbacks associated with conventional telescope receivers. As used herein, "transmission through free space" and "free space transmission" refers to a beam of radiation traversing an optical path, at least a portion of which is free of a waveguide, such as an optical fiber. Free space transmission, for example, includes transmission of radiation
15 through the earth's atmosphere such as between two earth-based stations or between an earth-based and space-based station and through space such as between two space-based stations. The receiver of the present invention can be much less inexpensive and smaller than conventional telescope receivers, and operated for long intervals of time and during daylight without damaging optical components in the receiver.

20 The laser communication receiver differs from previous methods of wavelength demultiplexing in that the present invention need not collimate, repoint and reroute to a separate demultiplexer signal power that is converging to a focus from the primary optical

lens. The detector units of the laser communication receiver of the present invention are in the converging beam of the primary optical lens. Since a diffractive optical element focuses different wavelengths to different on-axis locations, placement of detector units at these on-axis locations results in channel separation. The primary optical lens focuses at least most of the sunlight in front of beyond the leading detector unit. Only that portion, if any, of the sunlight having the proper wavelength is can be focused on each detector assembly.

The detector assembly can include a focusing element (in addition to the diffractive optical element) to reduce further the spot size of the radiation focused on the detector assembly. The focusing element can be any lens that further decreases the spot size of the radiation. Examples of such focusing elements include a solid immersion lens (SIL) and an oil immersion lens.

The additional focusing element is particularly effective in accounting for the detrimental effect not only of atmospheric turbulence but also of lens imperfections and window imperfections on data transmission. Imperfections in the primary optical lens can also smear out the focal spot. Because only small detectors, *e.g.*, 60 microns diameter, are able to respond to wideband, *e.g.*, 2.5 Gbps, data streams, the efficiency of any such detector unit is degraded by its inability to collect all the power in the focal spot. The focusing element increases detector efficiency by reducing the spot size of the focused radiation to a size smaller than the size of the photoactive area of the detector unit. The small spot size reduces the adverse effect not only of atmospheric turbulence but also of ~~and~~ lens imperfections on data transmission. The focusing element averages out such adverse effects by permitting the detector unit to collect a greater and more stable fraction of the radiation

contacting the receiver aperture than has previously been possible. This feature thus provides a more robust laser communication receiver.

The laser transmitter and laser receiver described above can provide low bit error rates for a given set of atmospheric conditions. Typically, the bit error rate is no more than about 1×10^{-9} and even more typically no more than about 1×10^{-12} . This is much lower than the typical minimum achievable bit error rate in most applications, which is approximately 1×10^{-6} .

Additional advantages of the present invention will become readily apparent from the following discussion, particularly when taken together with the accompanying drawings.

BRIEF DESCRIPTION OF THE DRAWINGS

Fig. 1 is a block diagram of a laser communication system which communicates at high data rates through the atmosphere;

Fig. 2 depicts a technique for reducing the adverse effect of the atmospheric turbulence at the transmitter aperture;

Fig. 3 depicts the relationship between divergence angle and the optical receiver;

Fig. 4 depicts a technique for positioning the optical receiver aperture relative to the divergent optical signal;

Fig. 5 is a schematic view of an optical receiver according to an embodiment of the present invention;

Fig. 6A is a schematic view of focal cones for different wavelength radiation being focused by a reflective hologram; and

Fig. 6B is a schematic view of two focal cones for differing wavelength radiation being focused by a transmissive hologram;

Fig. 7 is a schematic view of an optical detector assembly according to an embodiment of the present invention;

5 Fig. 8 is a perspective view of a transmissive holographic unit according to an embodiment of the present invention;

Fig. 9 is a perspective view of an optical receiver according to an embodiment of the present invention;

Fig. 10 is a cross-sectional view along line 10-10 of the optical receiver of Fig. 9;

10 Fig. 11 is a perspective view of a detector barrel for a six channel optical receiver;

Fig. 12 is a cross-sectional view taken along line 12-12 of the detector barrel of Fig. 11;

Fig. 13 is a front view of a detector assembly according to an embodiment of the present invention;

15 Fig. 14 is a plan view of a substrate including a plurality of detector units and a corresponding plurality of lenses or essentially hemispherical surfaces; and

Fig. 15 is a cross-sectional view of a detector unit and immersion lens according to another embodiment of the present invention.

20 DETAILED DESCRIPTION

The architecture of the present invention controls the detrimental effects of atmospheric turbulence at the transmitter by using a beam having a smaller diameter than the

inner scale as measured at the transmit aperture and/or that is divergent and/or broad and/or a receive aperture that exceeds the Fresnel scale (which is defined by $\sqrt{\lambda \cdot Z}$ where λ is the optical wavelength and Z is the distance from the transmit aperture). (to average out any smaller diameter intensity variations (within the beam footprint)). The use a smaller diameter beam has a significantly reduced likelihood of passing through differently moving air currents and thereby experiencing break up. The use of the divergent beam reduces the likelihood that the beam footprint will wander off of the receive aperture. As will be appreciated, a collimated beam of diameter "d" and wavelength "w" maintains its diameter for the Rayleigh range. After the Rayleigh range, the collimated beam diverges at an angle defined by w/d. The diverging beam diameter at any distance "z" from the transmitter (but beyond the Rayleigh range) is referred to as the footprint of the beam and is defined by wz/d.

The architecture controls the detrimental effects of atmospheric turbulence at the receiver by using a diffractive optical element to provide a plurality of converging optical signals (having spatially offset focal points) corresponding to differing optical wavelengths and/or an immersed detector to at least substantially maximize the optical signal power received by the detector. An essentially hemispherical immersion lens is positioned between the diffractive optical element and the detector to increase substantially the irradiance of the optical signal on the photoactive area of the detector.

With reference to Fig. 1, a simplex embodiment of a laser communication system 100 is shown in block diagram form. The laser communication system 100 includes a laser communication transmitter 104 and a laser communication receiver 102. As will be

appreciated, a duplex system may be implemented by placing a transmitter and receiver at each end of the link.

The Laser Communication Transmitter

The laser communication transmitter 104 can include one or a number of optical transmitters 132. The optical transmitters 132 each transmit a modulated laser beam 112 having a different wavelength λ . In one embodiment, there are four modulated laser beams 112 having wavelengths near 1550 nanometers where each wavelength differs by approximately 4 nanometers. Each laser beam 112 has a footprint which is increased by the atmospheric distortion 108. The footprint is a cross-sectional diameter of the laser beam 112 at the receive terminal focal plane 106.

Data is modulated in one or more of intensity, frequency, and phase onto a laser light carrier and is demodulated later from the laser light carrier in order to transport data great distances. The present embodiment assumes that intensity modulation of the transmit laser is used. In the laser communication transmitter, an optical transmitter 132 modulates data on a laser light carrier to form a modulated laser beam 112. The laser beam 112 is modulated with data at a rate greater than about 1 gigabit/sec. and preferably, greater than about 2 gigabit/sec. In order for the laser beam 112 to reach the laser communication receiver 102, the beam 112 travels through atmospheric turbulence 108.

The atmospheric turbulence 108 along the transmission path causes wavefront distortion in modulated laser beams 112. In other words, atmospheric turbulence 108 from thermal currents bends the collimated or diverging laser beams 112 and/or fragments the

beam, thereby causing mutual interference or cancellation and so produces burst errors in the data stream.

To overcome or reduce the effect of atmospheric distortion 108 on the bit error rate (BER), the diameter of each of the modulated laser beams 112 at the point 133 (Figure 2) at which each beam exits its corresponding transmitter 132 (or the aperture of the transmitter) is preferably selected such that the diameter is less than the inner scale of the atmospheric turbulence (or thermal currents) at the respective exit point.

As will be appreciated, the inner scale is typically a function of the temperature of any adjacent thermally conductive surface(s), the ambient air temperature, the distance from the transmit aperture to the adjacent surface(s) and wind speed. The inner scale is commonly from about 5 to about 20% of the distance " D_T " (Figure 2) from the adjacent surface 137 from which the thermal currents are emanating. For example, if the aperture of the transmitter is located at a distance D_T of 200 mm above a rooftop, the inner scale would be around 20 mm. Generally, it is desirable to position the transmitter aperture far enough from an adjacent heat absorbing surface such that the inner scale is at least about 20 mm.

In selecting the desired diameter for the beam at the transmit aperture, it is preferable to select the worst case for the particular location and environment in which the transmitter aperture is to be positioned. A suitable safety factor can then be applied to the inner scale and/or beam diameter or the beam diameter otherwise reduced to provide a beam diameter which will be safely below the inner scale under "worst case" conditions. Typically, the diameter of the (Gaussian) transmit beam 112 is no more than about 12.5 mm, more typically it ranges from about 1 to about 10 mm, and even more typically from about 2 to about 5 mm.

As will be appreciated, the beam diameter selection for each of the beams 112 is made independently. Each of the beams 112 could have a different diameter depending upon the magnitude of the inner scale at the point at which the beam 112 exits the corresponding optical transmitter 132.

5 The divergence angle of the broad beam is selected preferably by determining the “worst case” atmospheric induced deviation of the beam along the beam path and selecting the divergence angle of the beam (Figure 3) to exceed the turbulence-induced beam deviation for the beam under the worst case conditions. A suitable safety factor can be applied to the divergence angle, as desired. The divergence angle causes the far-field footprint of the beam 10
112 to remain on the receiver aperture despite atmospheric turbulence. Accordingly, beam 112 should be aimed properly at the receiver even when the beam is subjected to substantial deviation due to atmospheric turbulence 108. Typically, the divergence angle is at least about $20\mu r$ but needs be no more than about $2000\mu r$ and more typically ranges from about $50\mu r$ to about $1000\mu r$.

15 To provide further assurance that the far-field footprint of the beam remains on the face of the receiver, the receiver should be oriented properly relative to the incoming beam 112. Referring to Figure 4, the aperture 116 of the laser communication receiver 102 is preferably positioned/sized relative to the laser communication transmitter 104 such that the receiver 102 subtends at least about $20\mu r$ and more preferably at least about $50\mu r$ of the
20 diverging transmit beam 112 (as shown in Fig. 4).

Although Figure 1 illustrates a simplex (one-way) system, it is to be understood that the various inventions discussed herein can be used in a full duplex (two-way) system in which one or more transmitters and receivers are positioned at each end of the link.

The Laser Communication Receiver

Referring again to Figure 1, included in the communication receiver 102 is a holographic unit 116 (or other diffractive optical element), which is a primary optic for the laser communication receiver 102 and a detector assembly 106. The detector assembly 106 includes a focusing element 124 and a detector unit 128 for each modulated laser beam 112. In this embodiment, the transmitter 104 is located in a satellite and the receiver 102 is located in a ground station. However, other embodiments could put the receiver on a building and the transmitter on an airplane (because the receiver benefits are best realized when it is in atmospheric turbulence).

In the laser communication receiver 102, the holographic unit 116 focuses the laser beam 112 into a focus area on a focusing element 124 and detector unit 128. Referring to Figures 5, a single holographic unit 116 focuses each laser beam 112 on its respective focusing element 124 and detector unit 128, which is positionally offset (spaced apart) from the other (adjacent) focusing element 124 and detector unit 128. Typically, the distance D_s between adjacent detector assemblies 106a,b ranges from about 1 to about 10 mm with about 2 mm being more typical. As will be appreciated, Figure 5 is not drawn to scale, with the distance D_s being exaggerated relative to the diameter of the diffractive optical element 116. for clarity As can be seen from Figure 5, the longitudinal axes of the detector units 124 are typically parallel and collinear, and the longitudinal axes of the focusing elements 124 are

also typically parallel and collinear. The axes of these components are typically parallel to and collinear with the optical axis 138 of the holographic unit 116. As will be appreciated, the aperture size "A_s" of the holographic unit is determined by exceeding the Fresnel scale. By way of example, for a distance between the laser transmitter aperture and laser receiver aperture of 7.25 km and a wavelength of 1.55 microns, the Fresnel scale is $\sqrt{7250m \times 1.55\mu m}$ or 106 mm. The receive aperture is therefore larger than 106 mm and preferably about 400 mm. The modulated laser beams 112 are aimed at the holographic unit 116, whereafter, the laser beams 112 are focused by the holographic unit 116.

The operational differences between reflective and transmissive diffractive optical elements can lead to different receiver configurations. In reflective diffractive optical elements, the focused beam comes from the radiation-receiving side of the element while in transmissive diffractive optical elements the focused beam comes from the opposite side of the element from the radiation-receiving side.

Referring to Fig. 6A, for example, the operation of a reflective hologram 400 is depicted. The reflective hologram 400 focuses radiation 404, 408, corresponding to λ_1 and λ_2 respectively, at differing spacial locations along the optical axis 412 of the hologram. The hologram 400 includes a plurality of concentrically disposed, concave, spherical shells 414 in an emulsion. The locations 418 where the concave spherical shells 414 intersect the hologram surface 424 are spaced inversely with radial position relative to the optical axis 412 of the hologram. This feature ensures that different rings of radiation will phase up to form a cone of light converging to a focus 428, 432 (which correspond respectively to radiation

404 and 408) and that the distance of that focus from the hologram surface 424 will vary inversely with the wavelength of the radiation. For example, the distance of focus 428 from surface 424 is an inverse function of wavelength λ_1 , and the distance of focus 432 from surface 424 is an inverse function of wavelength λ_2 . A "thick" or "volume" hologram such as that in Fig. 6A has a number of weakly reflecting spherical shells 414 which will cause only radiation for which the spacing of the shells is about one-half of the radiation wavelength to be focused. All other wavelengths of radiation will pass through the hologram. A reflective zone plate would be formed by, for example, evaporating a plurality of concentric silver rings on the plate surface with the spacing between the rings (not shown) varying inversely with the radial position of the ring. The rings would be in the same positions as the ring-shaped intersections between shells 414 and hologram surface 424.

Referring now to Fig. 6B, a transmissive hologram 500 is depicted. The transmissive hologram 500 causes radiation 504 and 508 having wavelengths λ_1 and λ_2 , respectively, to focus at different spacial locations 528, 532 along the optical axis 512 of the hologram 500. The transmissive hologram 500 focuses light to the far side of the hologram. The hologram 500 includes a number of annular rings 514 (which appear as a plurality of concentrically disposed rings on hologram surface 524) having a spacing that scales inversely with radial position relative to the optical axis 512. This feature insures that at least some of the concentric rings of radiation passing through the hologram will phase up to create a converging cone of light. The hologram's focal length varies inversely with radiation wavelength. In this manner, all wavelengths shorter than 1.5 microns focus farther from the hologram than does 1.5 microns wavelength. Also, since the annular variations are difficult

to reproduce correctly near the optical axis 512 where the spacing is large, the central zone is typically useless. If concentric rings 514 were simply painted onto a flat glass plate, the hologram 500 would be a transmissive zone plate. Alternatively, the rings 514 can be formed by simply etching grooves on the surface. If, as shown, the rings 514 are index ripples and in emulsion, the optical element would be a hologram. As will be appreciated, other wavelength bands that are supported by available or developed transmit/receive technologies can also be used with the inventions set forth herein..

Although only holograms are discussed in the detailed description, it is to be expressly understood that any type of diffractive optical element can be employed in place of the holographic unit 116.

Referring again to Fig. 5, the holographic unit 116 is a transmissive hologram that preferably focuses each of the laser beams 112 into a converging cone. Because, as noted above, each laser beam 112 has a different wavelength λ , the focal area (i.e., the tip of the each cone-shaped focus) or focal point is at a different distance from the holographic unit 116. In this embodiment, the focal cone defined by lines 318a for λ_1 is about 60° to provide good separation between the modulated laser beams 112. Preferably, the focal cone for the various wavelengths ranges from about 30° to about 120° for optimum results.

As will be appreciated, the beams can contact the holographic unit simultaneously or sequentially (at different times), depending on the application. It is also possible for a beam to include multiple wavelengths of radiation, each wavelength representing a different channel and including different transmitted data.

As will be appreciated, the transmissive holographic unit 116 of Fig. 5 has a central obscuration 316 which blocks radiation. As noted, the obscuration is a result of the difficulty in producing the annular variations near the optical axis where the inter-annular variation spacing is large. The obscuration thus creates a shadow. To avoid optical losses, a first detector assembly is located in the shadow of radiation having shorter wavelengths that is detected by detector units positioned behind (or downstream of) the first detector assembly. Referring to Figure 5 for example, first radiation 318a of a first (mean or median) wavelength λ_1 is focused on a first focal area 320a at which a first detector assembly 106a is positioned. The set of dotted lines 324a represent the shadow of the central obscuration 316 for the first wavelength radiation but the shadow typically vanishes at the associated detector. Second radiation 318b of a second (mean or median) wavelength λ_2 is focused on a second focal area 320b at which a second detector assembly 106b is positioned. The set of dotted lines 324b represent the shadow of the central obscuration 316 for the second wavelength radiation. As can be seen, the first detector assembly 106a is positioned in the shadow of the set of dotted lines 324b and therefore the first detector assembly 106a does not cause significant optical losses of the second radiation. This effect can be realized by selecting proper wavelengths of radiation for use in each of the channels, such as for example wavelengths of 1550, 1554, and 1558 nanometers for a three-channel system.

After passing through the atmospheric distortion 108 and being focused by the holographic unit 116, the spot size at the tip of the cone-shaped focus of this embodiment is typically greater than about 100 microns or is approximately 200 microns in diameter. If a 40 micron detector were used with a spot size of 200 microns, only 4% of power would reach

the detector. The holographic unit 116 is a passive element and does not allow reshaping of the hologram surface to improve focus. Adaptive optics which allow reshaping of the optics can reduce this spot size with resultant improved focus, but they are expensive and complex.

Referring to Figure 7, the focusing element 124 overcomes the large focal area of the hologram by receiving the laser beam 112 from the holographic unit 116 and further focusing the light to reduce the spot size diameter. Reducing the spot size allows focusing more light while using a detector diameter small enough for compatibility with high speed transmission on the detector unit 128 which improves data transmission efficiency. The focusing element 124 can be any lens or optical component that increases the irradiance of the focal spot of the radiation. As will be appreciated, light can be focused to a spot whose diameter is inversely proportional to the refractive index of the lens medium 124. For example, a semiconductor immersion lens typically reduces the wavelength of light in air by a factor of about 3 to 4 (i.e., an index of about 3 to 4). In other words, the immersion lens reduces the wavelength of light while within the lens. Preferably, the focusing element increases the intensity of the light by at least a factor of nine. To realize these objectives, the element will typically have an index of refraction of at least about 2.3 and more typically that ranges from about 3.2 to about 4.0. The radius of curvature of the focusing element 124 typically ranges from about 400 to about 600 microns and more typically about 500 microns.

The focusing element 124 typically reduces the spot size of the incident radiation as focused by the holographic unit 116 to a diameter that is less than the photoactive diameter of the detector unit. In an exemplary application, the focusing element 124 reduces to 50 microns the typically greater than about 100 microns and more typically greater than about

200 microns spot size of the beam. Preferably, the spot size is reduced to about 60 microns or less and more preferably to about 40 microns or less. In other words, the spot size of the beam contacting the element 124 is preferably reduced by the focusing element the photoactive area of the detector unit or less.

5 The focusing element 124 is preferably one or more of a (solid) immersion lens, and/or an oil immersion lens. Preferably, the focusing element 124 is made of Silicon or Germanium which respectively have indices of refraction of 3.5 and 4 or InP or GaP with indices of refraction near 3.2. Silicon is preferred for wavelengths around 1500 nm and Germanium is preferred for wavelengths around 2000 nm to maximize throughput. Other possible SIL designs that can be employed in the present invention are described in U.S. Patent 6,114,688, which is incorporated herein by this reference.

10 The focusing element 124 is preferably in physical contact with the detector unit for best results. This physical contacting is known as "immersion". Immersion can be realized by any suitable technique, such as coaxial alignment and spring force or integral manufacture as discussed below.

15 The detector unit 128 receives and converts the modulated laser beam 112 into an electrical signal which contains the modulation data. At data rates above 1 gigabit/sec., wide band detectors are required, such as a PIN diode, avalanche photodiode or the like. However, wide band detectors with a large diameter are currently unavailable. As detectors get larger their capacitance increases. If large enough, this capacitance obscures the modulated data. Accordingly, for data rates above 1 gigabit/sec., currently available PIN diodes are generally limited to a diameter of no greater than approximately 60 microns.

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Referring to Figure 8, a transmissive holographic unit 116 according to another embodiment is schematically shown. The holographic unit 116 includes an emulsion layer 300, a flat glass disc 304, and a phase retarder 308.

The phase retarder 308 is any suitable material for altering the phase of the data pulse passing through the phase retarder, *e.g.*, a glass plate. Radiation that contacts the outer peripheral portion of the holographic unit (the area between radii r_1 and r_2) travels a further distance than radiation contacting the central portion of the holographic unit (the area having the outer radii r_1). This delay will cause the portions of the data pulse contacting in these areas to arrive at the detector(s) at different times, leading to smearing of data from differing radiation pulses. The phase retarder 308 causes the data from a given pulse of radiation to arrive at the detector(s) at more nearly the same time. The value of r_1 , or the location of the outer periphery 312 of the phase retarder 308, is based upon the desired maximum value of the predetermined time difference. The thickness "t" of the phase retarder typically ranges from about 10 mm to about 30 mm, and r_1 from about 50 to about 80% mm of the hologram radius.

In the absence of the phase retarder, radiation that is focused by peripheral portions of the primary optic will travel a longer distance than radiation that is focused by central portions of the primary optical lens. The different distances traveled can cause data from different pulses to arrive at overlapping times and therefore limit the usable modulation bandwidth. The phase retarder is typically positioned over the central portion of the primary optical lens to cause all portions of the same pulse to arrive within a predetermined time interval of one another and thereby avoid juxtaposition of data from differing pulses.

Referring to Figures 9-10, a preferred configuration of a six channel laser communication receiver is depicted. The receiver 600 includes transmissive holographic unit 116, a detector barrel 604 positioned in the central obscuration of the holographic unit 116, frustoconical outer housing 608, reflective surface 612, and phase retarder 308. The detector barrel 604 includes a plurality of detector assemblies 106a-f, each attached to a transparent substrate 616a-f; outer cylindrical housing 620, spacer rings 624a-e positioned between adjacent detector substrates 616a-f to maintain desired spacings between the detector assemblies 106a-f, end spacer ring 628 to maintain the rearmost substrate 616a in the desired position, locating flange 632 to ensure that the detector barrel 604 is mounted and retained in the proper position, and a plurality of coax connectors 636a-c (six total), each of which corresponds to a separate receive channel. The coax connectors 636 may be attached to a respective detector unit via conductors 640 located along the inner housing wall.

The reflective surface 612, as shown by dotted lines 644, causes incoming radiation to be directed in a reverse direction (or folded back) into the detector barrel. The reflective surface 612 is positioned at a distance "D" from the holographic unit 116 that is approximately equal to one-half of the focal length of the holographic unit 116. In an F/1 holographic unit, for example, the distance "D" is one-half of the aperture size of the holographic unit.

As will be appreciated, the reflective surface 612 could be omitted in which case the detector barrel 620 would be located in the base 648 of the housing 608. This design may not be preferred due to the significantly longer length of the housing, which can be

problematic in many space restricted applications. The reflective surface can be any suitable reflective optical component, such as a full silvered mirror.

Figures 11 and 12 depict a detector barrel 700 according to another embodiment. In contrast to the barrel 604, the barrel 700 includes a cylindrical inner detector ring 704 to position the rearward most detector substrate 616 in position. Spacer rings 620 are used as noted to hold the remaining detector substrates 616 in position. This barrel design further includes a length of a waveguide 708, such as a coax conductor that extends from each coax connector 636 to a respective detector assembly. As will be appreciated, the waveguide could be any other suitable RF interface to post detector electronics. The locating flange 712 is ring-shaped. The interface from the detector assemblies to the next stage of electronics (e.g. amplifier) can be any interface known to one of ordinary skill in the art.

Figure 13 depicts a preferred design for each detector assembly 106. The detector assembly 106 includes a detector unit 128 positioned on the substrate 616 which is transparent to radiation of the desired optical wavelength (Figure 10) and which is in the plane of the page. The typically square area 800 is the curved face of immersion lens or focusing element and the round area 804 is the photo active area of the detector unit 128. The width and height of the rectangular immersion lens typically each range from about 100 to about 300 microns and is more typically about 200 microns while the diameter of the round active area 804 typically ranges from about 30 to about 80 microns.

Because of the potential data losses in the conductors from the detector unit to the coax connector 636 at the high data transmission speeds noted above, the conductors are designed according to known transmission line theory. As will be appreciated, the

conductors are impedance matched with the detector unit and next stage of the electronics to reduce RF losses. The next stage is most likely an RF interface to a printed circuit board which is followed by an amplifier. The shape of the waveguide typically has stub sections to balance the photodiode impedance and may not be a pure rectangle as shown in the figure.

5 The transmission line can be designed "downward" from the detector unit. The transmission line is preferably a waveguide such as a microstrip or coplanar configuration. In one configuration, a microstrip transmission line 808 extends downwardly from the detector unit 128. The microstrip transmission line 808 includes a conductor 812 and ground plane 816. A first conductive material or "bump" 816, such as indium, attaches to the conductor 812 and
10 a second conductive material or "bump" 820 attaches to the ground plane 816.

The dimensions of the microstrip transmission line 808 depend on the application. By way of illustration, the ground plane 816 can be about 250 microns in width (" W_G "); the conductor 812 about 50 microns in width (" W_C "), and the conductor and ground plane separated by about 25 microns (or about 50% of the width of the conductor) of an electrically
15 insulative material, such as a polyimide. The length (" L_C ") of the conductors depends upon the pulse length of the beam and the impedance of the detector unit and next stage of electronics. For example, if the beam has a pulse length of about 4 inches the length " L_C " of the conductors should be less than 10% of the pulse length or less than about 0.4 inches. The cone of radiation 828 converging on the next detector unit is approximately 2 mm in
20 diameter in the plane of the detector unit 128 (or in the plane of the page) and is therefore not blocked to an appreciable extent by the detector unit 128.

The process for manufacturing the detector unit and the focusing element (which further focuses the focal spot of the diffractive optical element) will now be discussed with reference to Fig. 14.

Numerous problems exist in aligning the detector active area and the focusing element using conventional techniques. Traditional techniques of aligning solid immersion lenses with detectors include coaxial alignment and spring force. Traditionally such detectors were on the order of 3 mm in diameter and the immersion lenses of 12 mm in thickness. For acceptable performance in high data rate optical communication, detector sizes are much smaller (e.g., 60 microns) to respond fast enough to temporally resolve the 400 picosecond or shorter pulses of a 2.5 gigabit per second data stream. Not only is it difficult to fabricate the 200 micron solid immersion lens desirable for the 60 micron detectors but also it is difficult to properly orient the lens' axis to that of the much smaller detector unit and subsequently maintain the force needed to produce physical contact without breaking either component. To make matters worse, detector units will typically use a gold wire bond in the electrical connections. These wire bonds are at a minimum 25 microns tall and can interfere with the process of producing physical contact between the SIL and the detector unit unless the back side of the SIL is no larger than the detector unit.

To overcome these problems, a plurality of detector units 1004 are simultaneously formed on a transparent substrate 1000 by known techniques (e.g., epitaxial growth followed by photolithography), the substrate 1000 thereafter thinned, and on the opposite side of the substrate 1000, an essentially hemispherical surface 1008 is aligned and formed opposite to the detector unit 128. Typically, the optical axis of the surface 1008 is aligned (collinear)

with and is parallel to the center axis of the corresponding detector unit. In this manner, the alignment and attachment problems encountered with positioning a discrete immersion lens over a detector unit are avoided.

5 The substrate on which the detector(s) is grown can be any material that is transparent or semitransparent to the desired radiation wavelengths and has the properties referred to above for the focusing element. Preferred substrates include single crystals of indium phosphide, gallium arsenide, or gallium phosphide. Such substrates are transparent to incident signal power in the 1550 nm wavelength band. This combination of properties makes possible the admission of signal from the back (substrate) side through the substrate onto the detector unit.

10 After growth of the detector crystals 1004 and photolithography, the thickness of the substrate 1000 is typically thinned to the desired diameter of the focal element. For an f/1 lens, for example, the appropriate distance for a 200 micron diameter lens is also 200 microns, which is the reduced thickness of the substrate. Typically, the substrate 1000 is thinned to a thickness ranging from about 100 to about 400 microns.

15 The substrate 1000 is then etched by suitable techniques to form the plurality of focal elements or essentially hemispherical surfaces 1008, each of which is accurately registered or aligned to the location of a corresponding detector. Infrared microscopy can be used to align the mask used to deposit, for example, a photoresist mask on the substrate so that
20 lenses can be wet-etched. Alternatively, the etching can be performed by other techniques such as reactive ion etching, in which an ion beam is directed at individual regions of the substrate for appropriate lengths of time to remove sufficient material to form continuously

curved lenses. Another approach is to etch, either by wet process or by reactive ions, Fresnel-type structures as surrogates for the continuously curved lenses, or to approximate the continuous curves by a binary process which produces concentric steps at specific radii and height relative to reference axes. In addition to batch production of immersion lenses and precise coaxial registration of the lenses to a corresponding detector unit, the approach provides the intimate physical contact or detector immersion desired between the lenses and the detector units because they may be both fabricated onto the same substrate.

After batch processing is completed, the lens surfaces are anti-reflection coated. The anti-reflection coating can be any of a number of metal oxides or fluorides.

The substrate is then cleaved by any suitable technique (*e.g.*, or sawed) along horizontal and vertical cut lines 1012, 1016 (which are arranged in a grid pattern) to form a plurality of discrete detector assemblies, each including a substrate, detector unit, and hemispherical surface.

The detector unit can then be "bump" bonded, such as by indium, to another (second) transparent substrate on which the metal conductors have been previously deposited, such as in a microstrip or stripline configuration. Alternatively, the detector unit 128 could be positioned in a hole in the second transparent substrate on which conductors have already been deposited, as shown in Figure 10.

Another configuration for the detector assembly formed in this manner is shown in Figure 15. The assembly includes an immersion lens 900 in physical contact with the detector unit 904. A protective coating 908 is formed on the rear surface of the immersion lens 900 and the detector unit 904. Wire bonds 912 connect positive and negative leads 916

and 918 to the detector assembly 106. The immersion lens 900 is, as noted, formed by etching a GaAs or similar substrate on which the detector unit is grown. The detector unit is preferably formed from InGaAs and has a photoactive area ranging from about 30 to about 80 microns in diameter. The diameter of the immersion lens 900 preferably ranges from about 200 to 300 microns.

In this configuration, such coupling through the substrate also avoids another problem of mating SILs to InGaAs-on GaAs detector units – namely the need for the refractive index to remain essentially constant (within a range of about 20% of one another) from the SIL to the detector unit. The reason this is a problem is that fabricators of such detector units typically add a protective coating over the photosensitive InGaAs skin atop the substrate. This protective coating unfortunately has a very low refractive index (typically about 1.5) compared to the SIL (typically about 3.5) and the photosensitive InGaAs skin (typically about 3.5) so the coating degrades passage of signal between the SIL and the InGaAs. By using the GaAs substrate as the SIL, this interface problem disappears. The roughly 200 micron diameter SIL/detector unit would be mounted on transparent substrate or plates 616 (Figure 13), *e.g.*, microscope slides, which also support transmission lens design for detector output to the electrical preamplifiers (not shown).

The foregoing discussion of the invention has been presented for purposes of illustration and description. Further, the description is not intended to limit the invention to the form disclosed herein. Consequently, variations and modifications commensurate with the above teachings, within the skill and knowledge of the relevant art, are within the scope of the present invention. By way of example only, the invention need not be limited to a

hologram as the primary optical element. Any optical element can be used that has zones or areas of varying transmission, reflection, and/or varying refractive index could alternatively be used. The present invention is intended to include other receiver designs not discussed above which would result from using a transmissive and/or reflective diffractive optical element. In the reflective configuration, the detector barrel would be positioned on the laser beam receiving side of the diffractive optical element as the various wavelengths would be reflected by the optical element and received by the spaced apart detectors. Alternatively, the various features noted above can be used separately as part of other transmitter or receiver configurations to realize numerous benefits. For example, the use of a small diameter, so divergent broad beam can be used with conventional optical receivers to provide significant reductions in burst error occurrence. The optical receiver can be used with a conventional large diameter narrow beam to provide reduced bit error rates. The various elements of the optical receiver can be used in conventional optical receiver designs to further provide enhanced benefits. For example, the use of a larger aperture receiver can provide enhanced benefits. A focusing element, such as the SIL described above, can be used in connection with an otherwise conventional optical receiver to provide reduced fades in the received power. The receiver discussed herein and/or a component thereof may be used in optical communication applications other than free space transmission. By way of example, the receiver and/or component thereof may be used in fiber optic applications, such as a demultiplexer or optical-to-electrical converter. In this application, the receiver would be positioned so as to receive the optical signal transmitted by an optical fiber cable. An optical fiber can be substituted for each detector unit to collect separately each particular wavelength

of radiation. An example of such an application is described in U.S. Patent 4,643,519, which is incorporated herein by this reference. Other techniques can be used to produce the focusing element. Examples are discussed in U.S. Patents 3,239,675; 4,629,892; 4,636,631; 4,797,179; 5,545,896; and 5,652,813, all of which are incorporated herein by this reference.

5 The detector assembly can be composed of discrete, interconnected components or integrated components. For example, the detector assembly can be an integrated photodetector chip, which would have an amplifier built on the same chip as the PIN detector unit. This design could alleviate some of the engineering problems associated with bleeding off the small RF signal from the detector unit. Because the detector unit feeds a microwave waveguide for
10 some distance before amplification, the signal can pick up additional noise from the other densely packed detector-outputs and other sources (e.g., power supply). The RF signal could be swamped by noise before it reaches the active amplifier. An amplifier and other electronic circuitry for clock and data recovery traditionally follow the detector unit. Alternatively, a MEM (microelectronic mechanism or micromachine) can be integrated into
15 the assembly. MEM's could be applied to a tracking mechanism/feature for orienting/aligning the receiver and emitter. The embodiments discussed hereinabove are further intended to explain the best mode known of practicing the inventions and to enable others skilled in the art to utilize the inventions in such, or in other embodiments and with the various modifications required by their particular application or uses of the inventions.
20 It is intended that the appended claims be construed to include alternative embodiments to the extent permitted by the prior art.